Sequential Circuit Testing

Sequential Circuits and Finite State Machines

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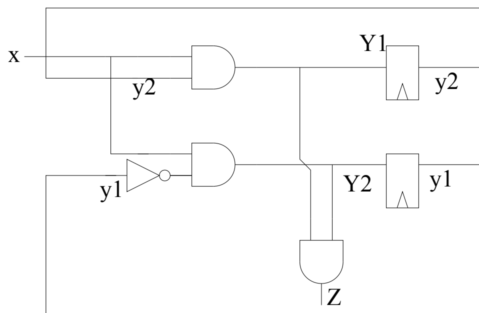
Slides updated Dec 1, 2021

Agenda

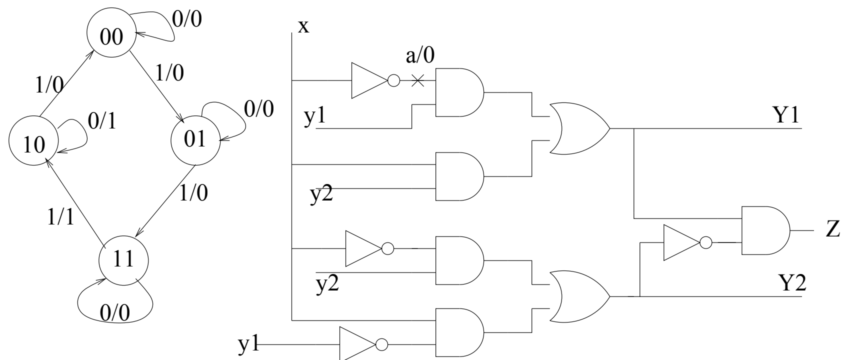
- Review of Sequential Circuits and Mealy Finite State Machines (FSM)
- The concept of FSM equivalence
- Fault excitation and propagation conditions in FSM
- Reset states, synchronizing sequences and redundant states
- Untestable faults in sequential circuits

Sequential Circuits

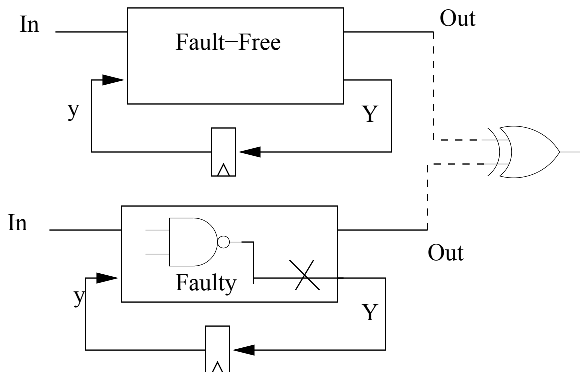
- We consider sequential circuits with edge-triggered D-flip-flops
- Underlying a sequential circuit is a finite state machine (FSM)
- We will consider only FSMs of the Mealy-type (Moore machine ATPG is similar)
- An example of sequential circuit:



FSM and Circuit example



Sequential Circuit Testing: Sequential Miter



- Find a **sequence of inputs** (or input vectors) that produces an output 1 at the miter output, sometime in the future clock cycles

FSM Equivalence

Two Machines, $\mathcal{M}_1$ and $\mathcal{M}_2$, are equivalent if

- They are identical; or
- They have identical states but different encoding; or
- $\mathcal{M}_1 \subseteq \mathcal{M}_2$ or vice-versa; or
- They have different reachable states but same distinguishable states (same condition as above); or
- Different unreachable states, and unreachable states are a *don't care* condition

Prove that two machines (sequential circuits) produce the same output response on application of all possible input **sequences**

Sequential ATPG is Harder than Combinational

- Given: A sequential circuit, generate ATPG. Sometimes reset (starting) state available, sometimes not!
- Procedure: Activate a fault, and propagate it to PO
- Bottlenecks: Fault activation requires an "activation state". How to get to the activation state? Fault can be propagated to the next-state line, but not to PO? Unroll the machine, or in other words, traverse to other states.
- How to distinguish between Faulty and Fault-free machines? The concept of state distinguishing seqences.
- In absence of resets, how to synchronize both faulty and fault-free machines to the same states. Concept of synchronizing sequences.

Untestable Faults in Sequential Circuits

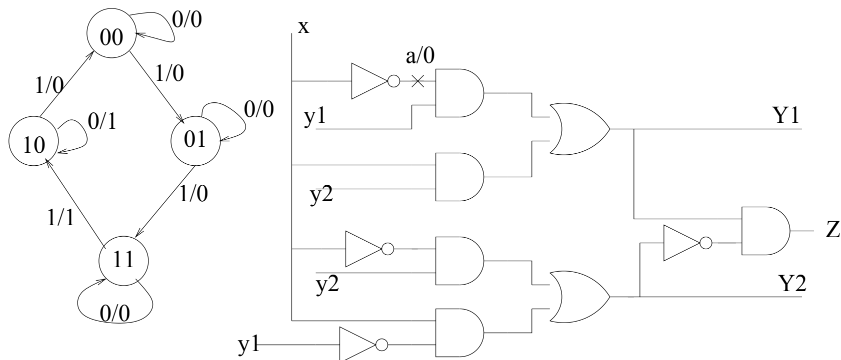
A fault in a sequential circuit can be untestable due to **four reasons**:

- Redundancy in a combinational logic
- Lack of a common synchronizing sequence that can drive both faulty and fault-free machines to a common (starting) state
- Sequential Redundancy: a fault causes a transition to a (faulty) state which may be equivalent to a fault-free state
- Fault excitation or propagation requires getting to an unreachable state

We will look at all these cases!

First example: When reset state exists

- Start the machine in a reset state (00), traverse FSM
- At some point in the future, faulty and fault-free machines will diverge



Test: $\underline{x_0 = 1}, x_1 = 1, x_2 = 0, x_3 = 1$

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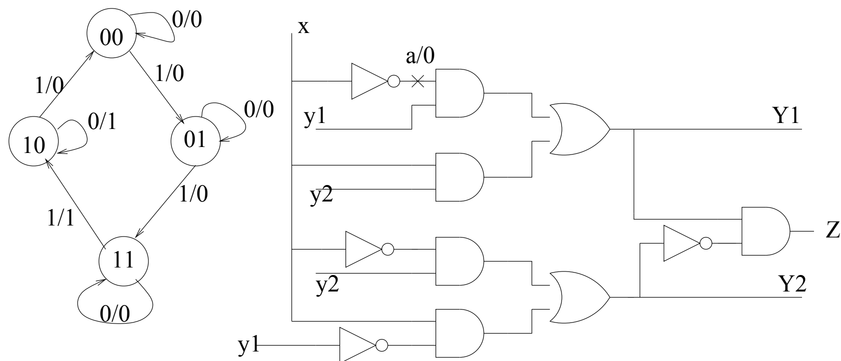

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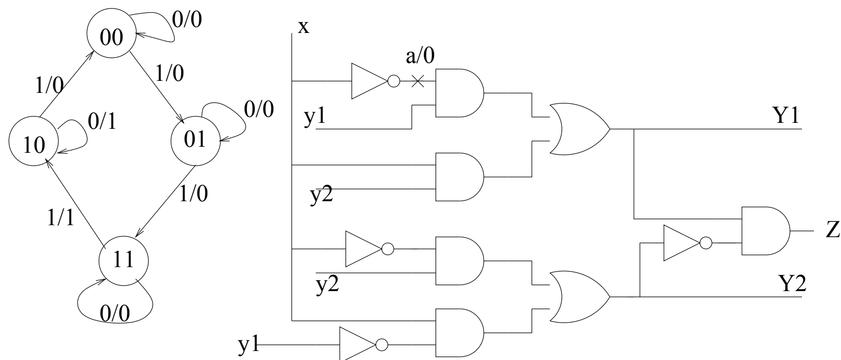

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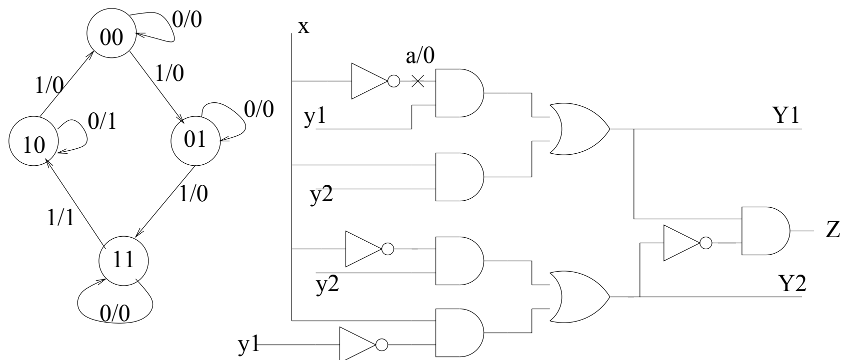


Test: $x_0 = 1, x_1 = 1, x_2 = 0, \underline{x_3 = 1}$

Meaning of the above Test

- Fault D *or* $\overline{D}$ gets into next state line.
- Faulty and Fault-free machines go to different states.
- Machine operation diverges....
- Sometime in the future, you can catch that effect.
- You can distinguish between faulty and faulty-free states of the machine.
- How? Using state distinguishing experiments.

When there is no reset state?

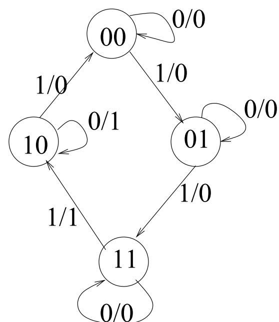
- Power up the machine, it can be in any (unknown) state
- Does there exist a sequence of inputs that can drive the machine to some – singleton – state?
- Some machines have a synchronizing sequence, others do not
- Given an FSM, how to find a synchronizing sequence?
- For ATPG: we are not given the FSM, only the circuit
- For ATPG: We need to find a synchronizing sequence for both the faulty and fault-free machines that can drive both to a common state. This can be hard.

Synchronizing Sequence on an FSM

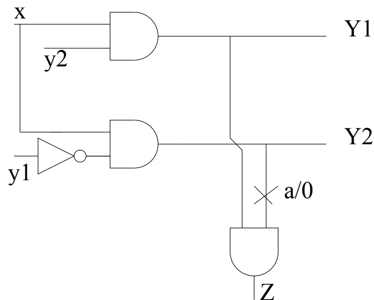
Table: State Transition Table

P.S.	Next State, Z	
	$x = 0$	$x = 1$
A	B, 0	D, 0
B	A, 0	B, 0
C	D,1	A,0
D	D,1	C,0

This machine has no Synchronizing Sequence



ATPG without Reset State



No Reset State Example: Test Exists

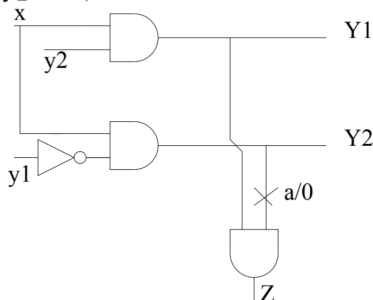
- No reset state. Start w/ Fault excitation state: $y_1 = 0, y_2 = 1$.
- $x = 1$ propagates the fault effect!
- Problem: How to get to the fault excitation state?
- Answer: Sequential backtrack! Unroll the m/c backwards until you can get unknown values in FFs. This implies there exists a state (backward in time from F.E. state) where you can "control" the FFs (control the state).
- Called Self-Initializing tests! This test implies that both faulty and fault-free machines can be initialized to a "common initial state"; i.e., you can get a common starting point.

ATPG without Reset State

Y = Next state FF inputs, y = present state FF output
Generate a self initializing test: Start in Time Frame (TF) 1:

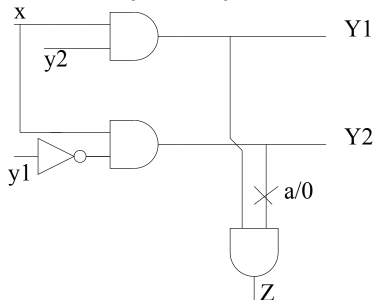
$$Y_2 = 1, Y_1 = 1, Z = D$$

This requires: $y_1 = 0, y_2 = 1, x = 1$



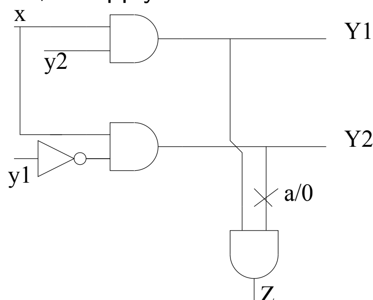
ATPG without Reset State

If in TF 1, present states are $y_1 = 0, y_2 = 1$, then in the previous state TF 0: $Y_1 = 0, Y_2 = 1$. Make $x = 1, y_1 = 0, y_2 = 0$



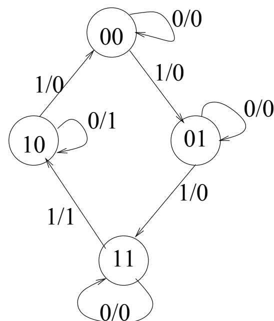
ATPG without Reset State

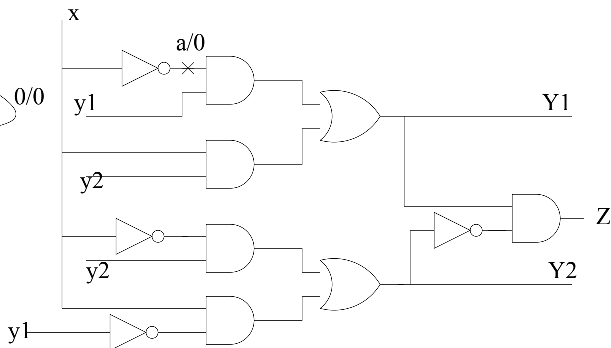
In TF 0 : $y_1 = 0, y_2 = 0$, then go back in time again TF -1:
$Y_1 = 0$, $Y_2 = 0$ is needed, so apply $x = 0$



Test: TF -1 $x = 0$, TF 0 $x = 1$, TF 1 $x = 1$.

No Self Initializing Test without Reset





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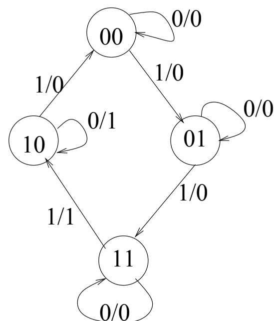
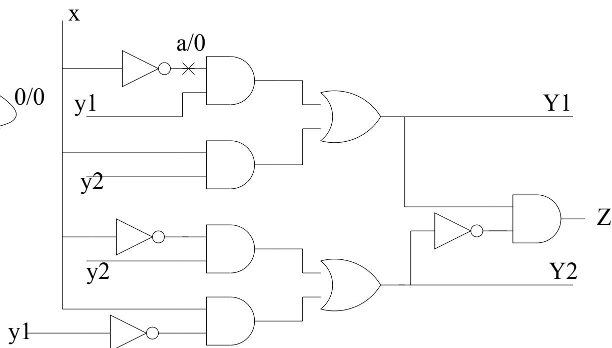
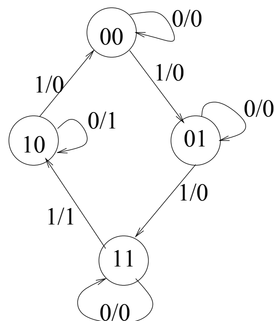
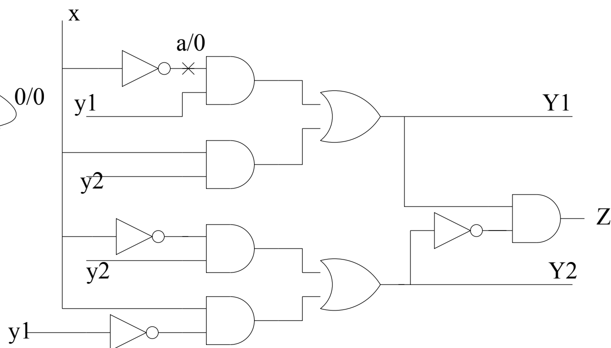




No Self Initializing Test without Reset





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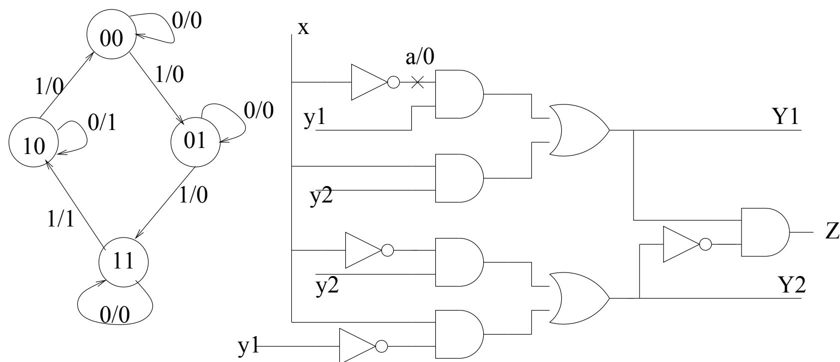
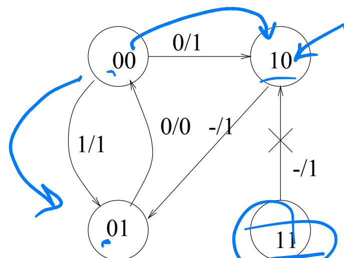


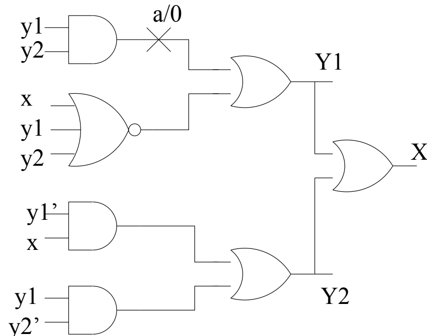
Therefore, no test exists, deduced after 4 clock cycles of sequential backtrace!

Untestable Faults: Unreachable State

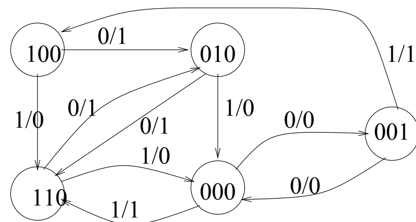


Y1 = y1 y2 +y1' y2' x'

Y2 = y1'x + y1 y2'



Untestable Fault: Sequential Redundancy

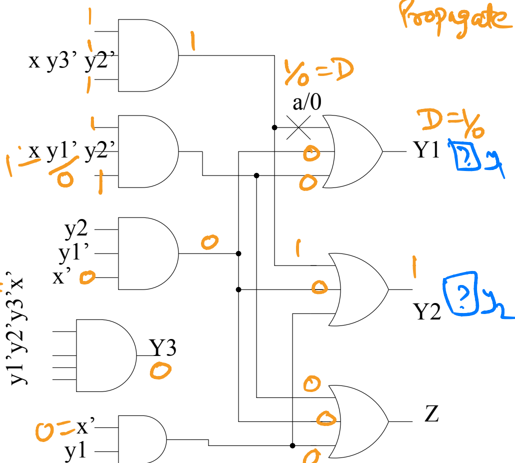


States (110) and (010) are equivalent -> cannot be distinguished!
What if fault takes m/c to 110 (faulty state), instead of 010 (fault free)?

Redundanct Circuit ATPG

Excite $x=1, y_3=0, y_2=0$

Propagate $y_1'=0 \Rightarrow y_1=1$



Excitation state

$y_1 y_2 y_3 = 100$

Next state

$Y_1 Y_2 Y_3 = D10$

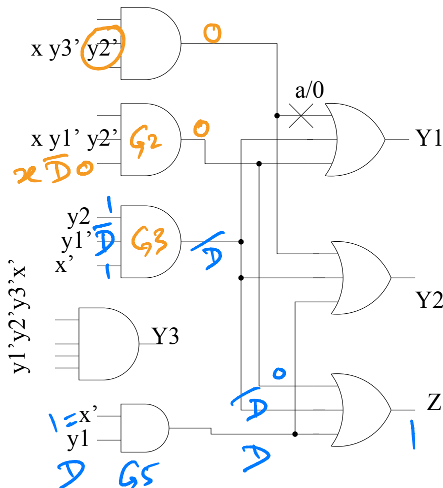
FF = 110

Faulty = 010

Z = 0

Redundanct Circuit ATPG

Next clk cycle $y_1 y_2 y_3 = PS = D10$



Propagate from $G_2 \to Z$

$y_2 = 1 \Rightarrow y_2' = 0$

$G_2 = 0.$

$G_3 \to Z$

$x' = 1 \Rightarrow x = 0$ reqd.

But $x' = 1$

$D + \bar{D} = 1$ @ Z

Fault masked!

Redundanct Circuit ATPG

x y3' y2'

x y1' y2'

y2

y1'

x'

y1'y2'y3'x'

Y3

x'

y1

a/0

Y1

Y2

Z

what about NS line @Y2?

$y_1 y_2 y_3 = D10$

$Y_2 = 1$

X

No test

Simplify Sequential ATPG using Scan Design

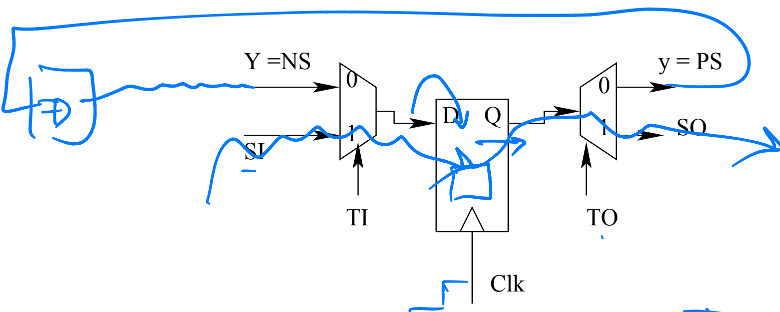
DFT = Design for Test

- Primary inputs = controllable, Primary outputs = observable
- Lack of controllability and observability of Flip-Flops
- For testing purposes, make the flip-flops both controllable and observable
- Approach: Introduce Scan-registers
- Make a sequential circuit combinational for Testing

LSSD

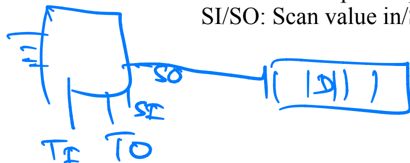
Scan Flip-Flop Design

normal mode TI = TO = 0

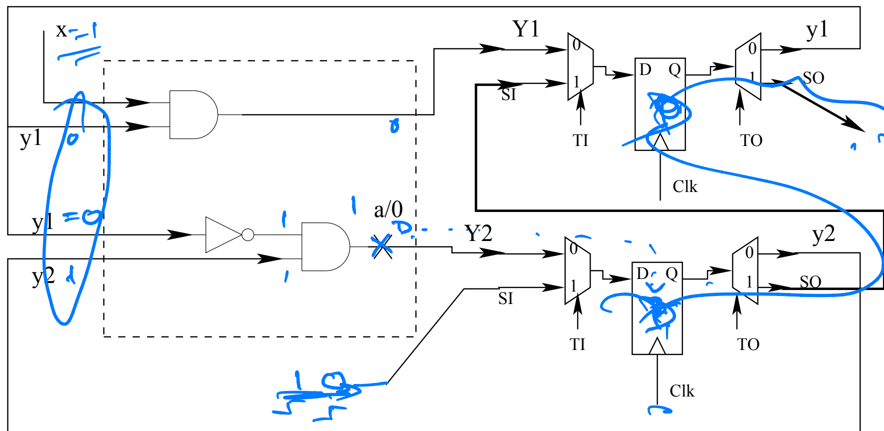


TI/TO: Test input/output
SI/SO: Scan value in/Scan value out

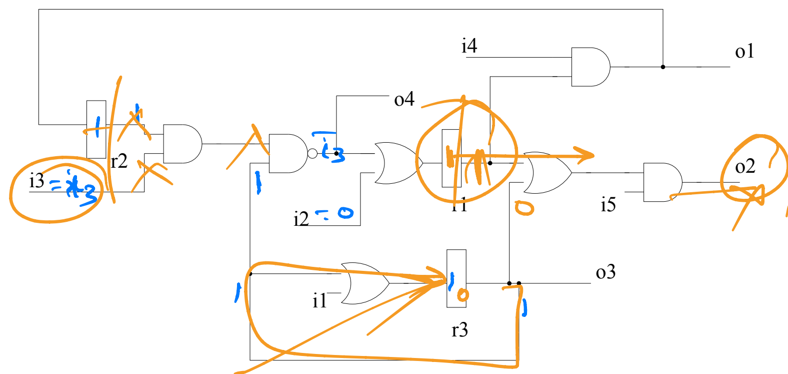
TI=1
TO=1



Scan Based ATPG: Use Shift-register Scan Chain

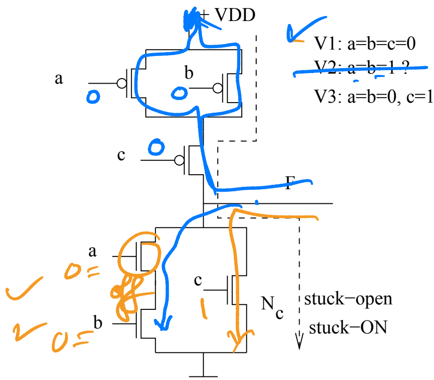


Other Test Problems: False Paths



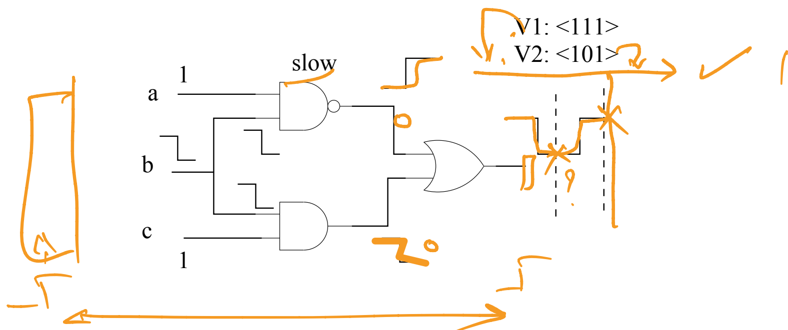
- Input $i_3 \rightarrow r_1 \rightarrow o_2$ is a multi-cycle False Path
- Cannot be sensitized: Need $o_3 = 1$ in clock cycle 1, and $o_3 = 0$ in clock cycle 2. Not possible

- Fault Location in Transistors: Test I_{DDQ}
- Transistor can be stuck-ON or stuck-open. CMOS gates comsume NO static power.
- Test for N_c stuck-open: $\langle V_1, V_3 \rangle$ Two vector test



BIST = built-in self-Test

- Gates do not switch at desired speed
- Find vectors that detect delay faults!
- Big challenge in Test Application: How to apply 2-vector tests using scan flip-flops at speed?



Delay Fault Testing

- Gates do not switch at desired speed
- Find vectors that detect delay faults!
- Big challenge in Test Application: How to apply 2-vector tests using scan flip-flops at speed?